



A Novel Approach to Reduce Area and Power for FFT Implementation

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Abstract: The Fast Fourier Transform (FFT) is a critical block widely used in digital signal processing algorithm. With the advent of semiconductor processing technology in VLSI system, different approaches had been tried in order to optimize the algorithm for a wide variety of parameters such as area, power and speed. In this paper, we propose to design a FFT block which is capable of computing N point FFT based on Radix-2 Decimation-In-Time (DIT) architecture with carry select adder (CSLA) and a gate level modification to CSLA. Here our goal is to implement Radix-2 N-point FFT in hardware using hardware language (VHDL). Simulation of design units is done in ModelSim-Altera Starter Edition and synthesized using Xilinx ISE 8.1. The overall area and power are reduced

Keywords: FFT; Carry Select Adder; Radix-2;VHDL

I. INTRODUCTION

To perform the frequency analysis on a discrete time signal, we convert the time domain sequence to an equivalent frequency domain representation. Such a representation is given by the Fourier transform. The discrete Fourier transform (DFT) is one of the Fourier transform, used for Fourier analysis. It transforms time domain signals to frequency domain signals [5]. In an N point DFT, to evaluate each DFT sample value, we have to perform N multiplications and N-1 additions using complex numbers. N such computations are required in all, therefore there will be N^2 complex multiplications and $N(N-1)$ complex additions.

The FFT procedure for synthesizing and analyzing the Fourier series was given by Cooley and Tukey. It is a computationally efficient way to calculate DFT. The wide usage of DFT's in Digital Signal Processing applications is the motivation to implement FFT's. This method provides a divide and conquer approach to the computation of DFT [2]. This is based on the decomposition of N point DFT into successively smaller DFTs [1]. In an N point sequence if N can be expressed as $N = r^m$, then the sequence can be designated into r point sequences where m denotes the number of stages of computation and r denotes radix of the FFT algorithm. For example to perform radix 2 FFT algorithm, the value of N should be such that $N = 2^m$.

Here the decimation can be performed 'm' times where $m = \log_2 N$. The total number of complex additions are reduced to $N \log_2 N$ and total number of complex multiplications are reduced to $(N/2) \log_2 N$ [1]. There are two methods of FFT: Decimation In Time (DIT), and Decimation In Frequency (DIF). In DIT, the inputs are fed in bit reverse order and the outputs are obtained in normal order. In DIF, the inputs are given in normal order to the butterfly unit and outputs are obtained in bit reversed format. This paper adopts Radix-2, decimation-in-time FFT algorithm [1].

II. RADIX-2 FFT ALGORITHM

Radix-2 is the first FFT algorithm and was proposed by Cooley and Tukey in 1965 [1]. The DFT of a given sequence $x[n]$ can be computed using the formula

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{nk}, \quad k=0,1,\dots,N-1 \quad (1)$$

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Vol. 2, Special Issue 1, December 2013

where $W_N = e^{-j2\pi/N}$ is the Twiddle factor which is referred to as roots of unity complex multiplicative constants in FFT computation[3]. They can be used to recursively combine smaller length DFTs. In radix-2 FFT, the N-point DFT can be decomposed into (N/2) point even and odd subsequences as[6]:

$$x_1(r) = x(2m) \tag{2}$$

$$x_2(r) = x(2m + 1); \quad m = 0, 1, \dots, (N/2) - 1 \tag{3}$$

Then, the N-point DFT becomes

$$\begin{aligned} X(K) &= \sum_{n(\text{even})=0}^{N-1} x(n)W_N^{nk} + \sum_{n(\text{odd})=0}^{N-1} x(n)W_N^{nk} \\ &= \sum_{m=0}^{(N/2)-1} x_1(2m)W_N^{nk} + W_N^{nk} \sum_{m=0}^{(N/2)-1} x_2(2m + 1)W_N^{2mk} \end{aligned} \tag{4}$$

The properties of twiddle factor used are: periodicity ($W_N^{k+N} = W_N^k$) and symmetry ($W_N^{k+N/2} = -W_N^k$). Then we get,

$$X(k) = X_1(k) + W_N^k X_2(k) \tag{5}$$

$$X(k+N/2) = X_1(k) - W_N^k X_2(k) \tag{6}$$

Where, $k = 0, 1, \dots, N/2 - 1$.

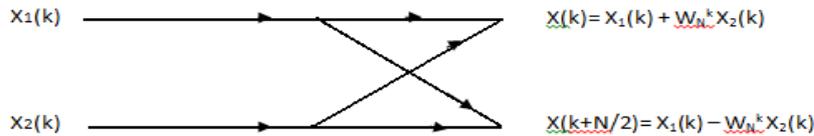


Fig.1 Flowgraph of basic butterfly unit

If N is a regular power of 2, the same computational procedure can be applied recursively until the N-point DFT is evaluated as a collection of 2- point DFT's[5].

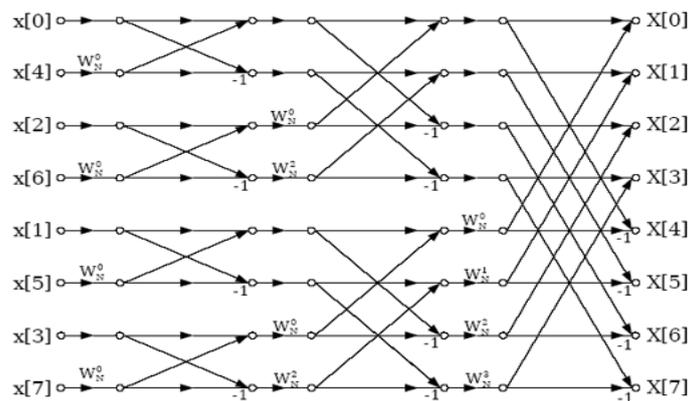


Fig.2 Flowgraph of 8-point Radix-2 DIT

III. FFT ARCHITECTURE

The block diagram of a general FFT architecture is shown in the figure 3. It consists of five basic blocks. A double port RAM memory to hold the values of input, output and intermediate operations, a Butterfly processing unit which consist of radix-2 butterflies and it is the heart of FFT algorithm, ROM memories to store twiddle factors and an address generation unit to extract data from RAM and ROM and finally a controller.

International Journal of Advanced Research in Electrical, Electronics and Instrumentation Engineering

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Vol. 2, Special Issue 1, December 2013

A. General Description

The whole FFT operation is partitioned into three processes: DATA LOAD, COMPUTATION AND RESULT UNLOAD. Initially input datas are given to the processor by means of siminputs. These datas are written into the RAM. The processing cycle starts with data load process. Here we read the stored data from RAM and start signal for FFT becomes high. Then, FFT computation of stored data takes place. Finally the results are stored in RAM and becomes available at the output[3].

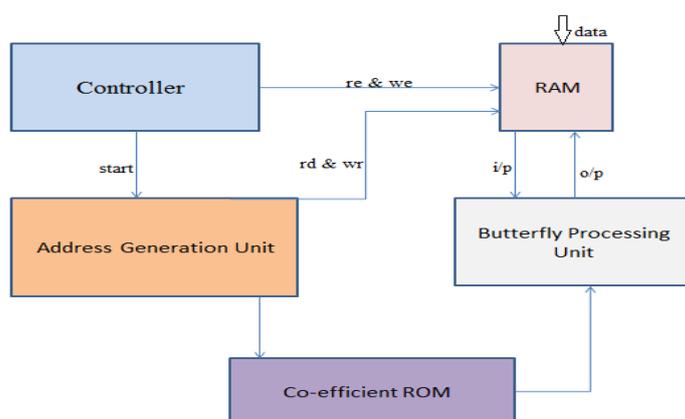


Fig. 3 Block diagram of FFT

B. RAM

The RAM memory carries different types of data as the computation of FFT algorithm proceeds[3].Initially the input datas for FFT computation are stored in RAM. After each butterfly operations,it overwrites the input data positions. And finally during the output process bit reversed address is given to the RAM and it outputs the data accordingly. The dual port capability of RAM makes two data samples available at the sametime.So it decreases the overall computation time of FFT.

C. ROM

The ROM is used to store the sine and cosine values of twiddle factors needed in the FFT computation process[3]. It outputs these values according to the address given to it.

D. Butterfly Processing Unit

The butterfly is the basic operator of the FFT. It takes two data words from memory and computes the two point FFT. The input data are taken in 2's complement format[3]. A butterfly unit consist of (N/2) butterflies and the basic structure is shown in the figure 4. Each unit contains two ROM memories to store sine and cosine co-efficients of twiddle factors, four 16x16 bit multipliers, six 32-bit accumulators and two concatenation operators to get the correct data format at the output.

The arithmetic operations involved in the butterfly unit is shown in the table I. The whole butterfly operation takes six instants of time. First, the two data inputs and the twiddle factor co-efficients are read(R).They are multiplied(x) to get the partial products in 16-bit format. Then, they are added and subtracted and truncated (+/-).The truncated real and imaginary parts are concatenated (&) to get the 32-bit format.

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The adder used in butterfly computation is the carry select Adder (CSLA) which is one of the fastest adder [9]. The main advantage is that the arithmetic operations follows pipelining operation [4],[8] and addition is done using CSLA, so it reduces the overall computation time.

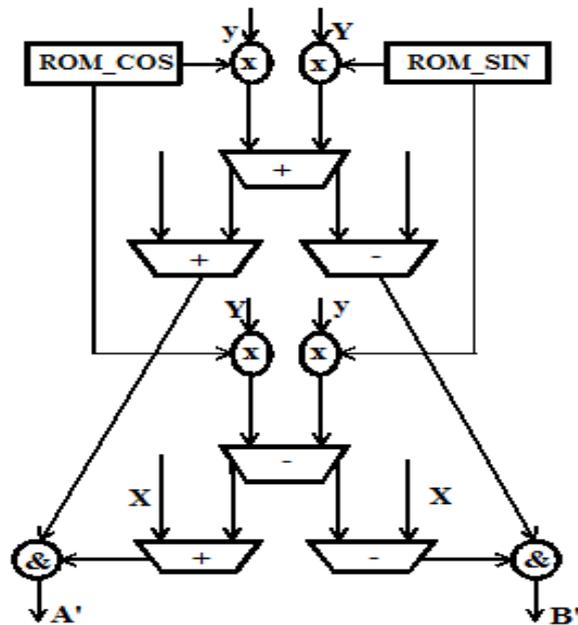


Fig.4 Butterfly Structure

Table1. Arithmetic operations in butterfly block

	1	2	3	4	5	6
R (Read data)	cos					
	sin					
	A					
	B					
x (Partial products)		$m1=y\cos$				
		$m2=Y\sin$				
		$m3=y\sin$				
		$m4=Y\cos$				
± (Terms grouping and truncation)			$s1=m1+m2$			
			$s2=m4-m3$			
				$s3=x+s1$		
				$s4=x-s1$		
& (Format Recovery)						
					$A'=s3\&s5$	
					$B'=s4\&s6$	
W (Write results in RAM)						A'
						B'

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E. Address Generator

The purpose of address generation unit is to provide the ROM and RAM memory with correct address. It also keeps track of which butterfly is being computed in which stage. The block diagram of this unit is given in figure 5. For an 8-point complex FFT, there are 3 stages, each stage consisting of 4 butterflies. Since the address during input, output and FFT computation processes are different, it keeps track of the mode of operation of the chip and generates the required address [3]. Mode of operation information is supplied by the controller.

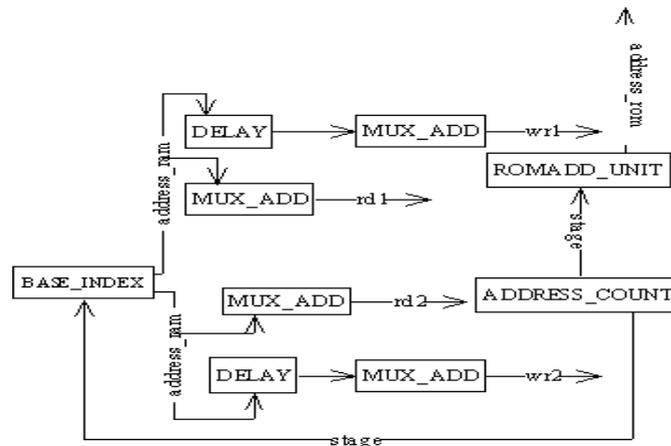


Fig.5 Address generation unit

F. Controller

It controls the whole activity of FFT computation. It act as a combinational state logic.

IV. PROPOSED ADDER IN BUTTERFLY UNIT

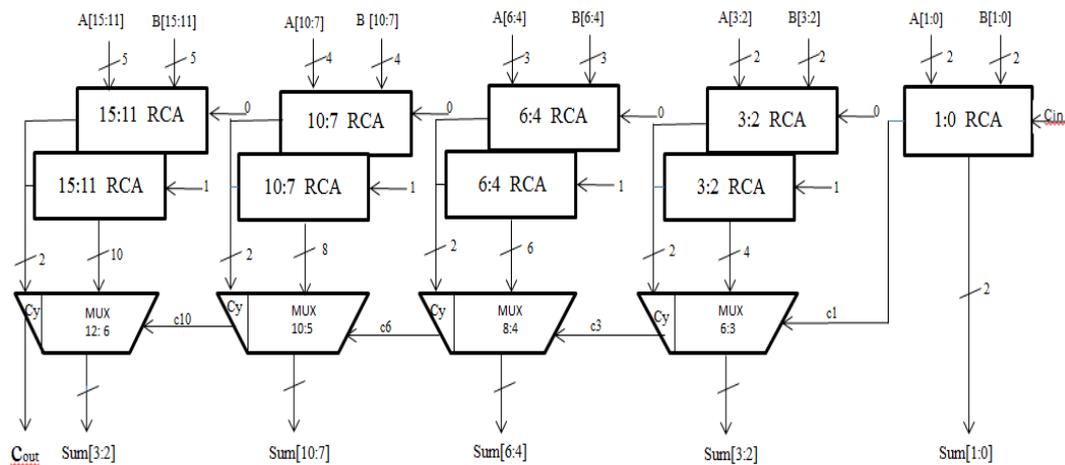


Fig.6 16-bit CSLA

The adder circuit used in the butterfly structure is the carry select adder which consists of two Ripple Carry Adders (RCA) with $C_{in}=0$ and $C_{in}=1$ and a Multiplexer [9]. The problem in CSLA design is that it is not area

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efficient because it uses multiple pairs of RCA to generate partial sum and carry by considering carry input and then the final sum and carry are selected by the mux which is shown in the figure 6.

In order to improve the shortcomings of CSLA, we use Binary to Excess-1 Converter (BEC) instead of RCA with $C_{in} = 1$ in the regular CSLA[9]. The structure of proposed adder is shown in the figure 7. The main advantage of this BEC logic comes from the lesser number of logic gates than the n-bit Full Adder(FA) structure. The modified CSLA architecture is therefore, having low area, low power, so it is simple and efficient for VLSI hardware implementation.

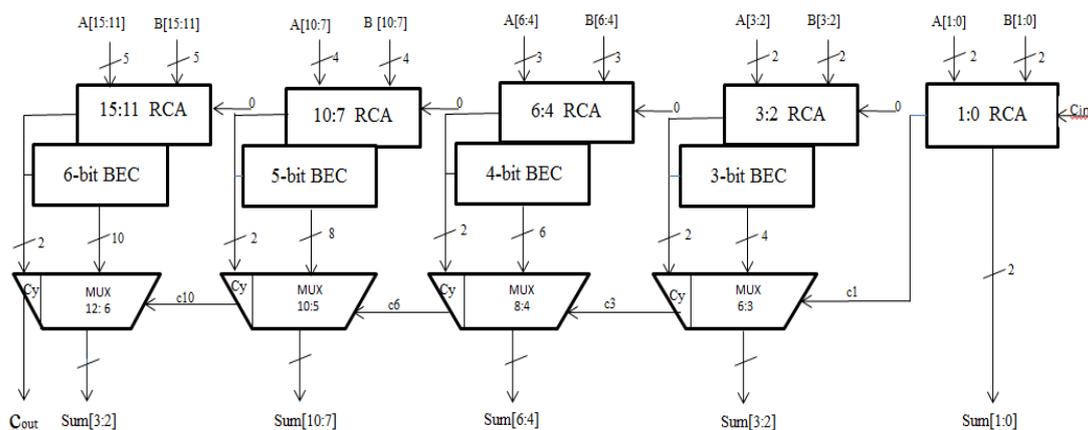


Fig.7 Modified CSLA with BEC

V. RESULTS AND DISCUSSIONS

After simulating both CSLA and proposed CSLA with BEC incorporated FFT, the design is synthesized using Xilinx ISE 8.1. A comparison between existing FFT and proposed FFT is done.

A. Comparison between CSLA and CSLA with BEC

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Device Utilization Summary				
Logic Utilization	Used	Available	Utilization	Note(s)
Total Number Slice Registers	668	12,288	5%	
Number used as Flip Flops	454			
Number used as Latches	214			
Number of 4 input LUTs	792	12,288	6%	
Logic Distribution				
Number of occupied Slices	543	6,144	8%	
Number of Slices containing only related logic	543	543	100%	
Number of Slices containing unrelated logic	0	543	0%	
Total Number of 4 input LUTs	792	12,288	6%	
Number of bonded IOBs	48	240	20%	
Number of BUFG/BUFGCTRLs	1	32	3%	
Number used as BUFs	1			
Number used as BUFGCTRLs	0			
Number of FIFO16/RAMB16s	6	48	12%	
Number used as FIFO16s	0			
Number used as RAMB16s	6			
Number of DSP48s	2	32	6%	
Total equivalent gate count for design	10,898			
Additional JTAG gate count for IOBs	2,304			

Fig.8 Design summary of regular CSLA incorporated FFT

Regular CSLA incorporated FFT is considered first. The device utilization summary of regular CSLA incorporated FFT is shown in the figure 8. Percentage of total number of Slice Registers and percentage of total number of 4 input LUTs is found to be 5% and 6% respectively. Also the total equivalent gate count for design is found to be a 10,898.

Logic Utilization	Used	Available	Utilization	Note(s)
Total Number Slice Registers	559	12,288	4%	
Number used as Flip Flops	454			
Number used as Latches	105			
Number of 4 input LUTs	707	12,288	5%	
Logic Distribution				
Number of occupied Slices	510	6,144	8%	
Number of Slices containing only related logic	510	510	100%	
Number of Slices containing unrelated logic	0	510	0%	
Total Number 4 input LUTs	717	12,288	5%	
Number used as logic	707			
Number used as a route-thru	10			
Number of bonded IOBs	48	240	20%	
Number of BUFG/BUFGCTRLs	1	32	3%	
Number used as BUFs	1			
Number used as BUFGCTRLs	0			
Number of FIFO16/RAMB16s	6	48	12%	
Number used as FIFO16s	0			
Number used as RAMB16s	6			
Number of DSP48s	2	32	6%	
Total equivalent gate count for design	9,789			

Fig.9 Design summary of modified CSLA incorporated FFT

The device utilization summary of Modified CSLA with BEC incorporated FFT is shown in the figure 9. Percentage of total number of Slice Registers and percentage of total number of 4 input LUTs is found to be 4% and 5% respectively. Also the total equivalent gate count for design is found to be a 9,789.

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Table 2. Synthesis results of both regular and modified CSLA incorporated FFT

Logic Utilization	Regular Carry Select Adder	Modified Carry Select Adder
Percentage of total number of Slice Registers	5%	4%
Percentage of total number of 4 input LUTs	6%	5%
Total equivalent gate count	10,898	9,789

Table III shows the power consumption of both regular and modified CSLA incorporated FFT. The overall consumption of power is reduced from 334mW to 177mW in modified CSLA with BEC incorporated FFT.

Table 3. Power Consumption

Power	Regular Carry Select Adder	Modified Carry Select Adder
Total estimated power	334mW	177mW

Thus it is evident that resource utilization and power consumption is reduced in latter. So the CSLA with BEC in the design of butterfly unit optimize the area and power of FFT core.

B. Simulation result of 256-point FFT

The design is developed in VHDL language and simulated using ModelSim-Altera 6.4a. For the implementation of this circuit we have used Xilinx FPGA (XC4VLX15 device of Virtex 4). The top level source is Hardware Description Language type, with a XST (VHDL/VERILOG) environment. The simulated waveform of 256-point FFT is shown in figure 10.



Fig.10 Simulated waveform of 256-point FFT

The inputs of FFT are taken in 2's complement format and the results are verified using MATLAB. The computation of FFT algorithm with different points can be simulated by varying the size of ROM and address generator



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unit while the size of other sub-blocks remains the same. This flexibility makes this FFT algorithm applicable for various re-configurable VLSI applications [7].

VI. CONCLUSION

The design of Radix-2 FFT with 256-point is done and simulated using Modelsim-Altera 6.4a Starter Edition. Our structure is synthesized in Xilinx ISE 8.1. A simple approach is proposed in this paper to reduce the area and power of FFT architecture. The reduced number of gates by replacing regular CSLA with modified CSLA with BEC offers the great advantage in the reduction of area and also the total power.

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